

A COMPACT DESIGN OF A BALANCED 1×4 OPTICAL POWER SPLITTER BASED ON SILICON ON INSULATOR SLOT WAVEGUIDES

M.A. Butt¹, A. Naresh Kumar Reddy¹, S.N. Khonina^{1,2}

¹ Samara National Research University, Samara, Russia,

² Image Processing Systems Institute of RAS – Branch of the FSRC “Crystallography and Photonics” RAS, Samara, Russia

Abstract

In this paper, a compact design of a balanced 1×4 optical power splitter based on coupled mode theory (CMT) is presented. The design consists of seven vertically slotted waveguides based on the silicon-on-insulator platform. The 1×4 OPS is modelled using commercial finite element method (FEM) simulation tool COMSOL Multiphysics 5.1. The optimized OPS is capable of working across the whole C-band with maximum ~39 % of power decay in the wavelength range 1530 – 1565 nm.

Keywords: slot waveguides, 1×4 power splitter, SOI, C-band, finite element method.

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Introduction

Research on silicon photonics can be dated back to the 1980s. Though, the previous decade has observed an extraordinary development in the field. Based on highly advance silicon semiconductor technology, silicon photonics would deliver with an inexpensive, highly integrated electronic-photonics platform, in which ultra-compact photonic devices and electronic circuits are converged. The fundamental motive behind silicon photonics is the ability to use CMOS-like fabrication resulting in high-volume production at low cost [1]. Optical power splitters (OPS) play a fundamental role in optical communication systems [2, 3] and periodic optical sources for integrated microfluidic devices [4]. These optical elements are substantial for conveying the optical fiber to end-users [5]. OPSs are principally based on either Y-branch splitters [3, 6, 7] or multimode interference (MMI) splitters [8–10]. Generally, the former is more compact but the optimization of their branch region is complicated. The latter is simpler, while it has large modal interaction length. For optical elements based on modal interference, reduction of the modal interaction length can result in increased sensitivity to fabrication tolerances and therefore can sacrifice device robustness. For instance, shortening the interaction length of directional couplers by reducing the gap between waveguides can affect the fabrication tolerances, which are already very low due to the comparatively tight guiding that is characteristic of semiconductor waveguides.

There are several designs of OPS, which are proposed recently. 1×2 and 1×3 Ultra compact power splitters based on the horizontal nano-plasmonic slot waveguides are proposed [11]. A 1×4 OPS based on multimode interference (MMI) coupler in silicon (Si)-gallium nitride (GaN) slot waveguide structure is presented. This device can be useful to divide optical signal energy uniformly in the C-band range into four output ports [12]. In [13] 1×4 OPS is proposed by combining three 1×2 OPS based on two-mode interference (TMI) effects in slot waveguides.

An exceptional waveguide geometry known as a slot waveguide is presented by Almeida et al in 2004 where the guided light is strongly confined within a narrow low-index slot (S) between the two high indices photonic wires [14, 15]. There are two claddings (high index material) separated from each other with a narrow slot (low index material) [16, 17]. The electric field propagating in the slot suffers a disruption at the high refractive index contrast interface which makes the electromagnetic wave to confine intensely in the narrow slot than in the cladding which can lead to the implementation of compact and high-performance photonic components such as resonators [18] and polarization beam splitters [19, 20]. Moreover, there are no confinement losses in the slot waveguide structure due to the strong high power confinement inside the slot area. Therefore, there is a noteworthy attention in designing photonic devices based on a slot waveguide structure that integrates semiconductor materials.

In this paper, we modelled 1×4 OPS of size 8 μm based on coupled-mode theory (CMT) [21] by using seven vertically slotted waveguides at 1550 nm TE-polarized light. Numerical optimizations were performed on the coupling gap (C1, C2 and C3) between the slot waveguides structures (by keeping the coupling lengths constant) in order to obtain an equal power distribution among 4 outputs. The CMT is used to evaluate the mode coupling or conversion in optical waveguides. The physical models for coupled waveguide systems contain two or more dielectric waveguides employed in close proximity. These waveguides maybe parallel to each other or may have variable separations. The slot waveguide core is composed of two Si ($n = 3.48$) slab surrounding a narrow slot. The slot is simply an air ($n = 1$). The splitter schematic structure is shown in fig. 1. And the slot waveguide cross section is depicted by the inset. All the simulations are performed by using Comsol Multiphysics 5.1 which solves the Helmholtz equation with the finite element method (FEM).

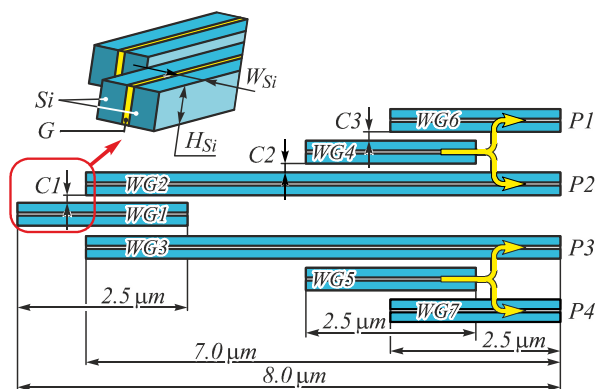


Fig. 1. Schematic of a 1×4 OPS

1×4 OPS model

In our proposed design, we used seven vertically slotted waveguides in order to form a compact 1×4 OPS. W_{Si} , H_{Si} and G are chosen at 150 nm, 220 nm and 40 nm respectively, in order to acquire a maximum confinement in the slot region and ensure a fundamental TE mode in the proposed power splitter. The WGs are placed in a close proximity so that the evanescent modes of one WG overlap with the modes of a second guide. The light from one WG gradually transfers back and forth between the WGs. It is possible to couple any desired fraction of light from one WG to other by choosing the separation between the WGs and length of the coupling region,

The length of WG1, WG4, WG5, WG6 and WG7 is fixed at 2.5 μm which provides an adequate coupling length between the adjacent WGs. While the length of the WG2 and WG3 is 7 μm which serves as output at port 2 and port 3. The output from WG6 and WG7 is collected at port 1 and port 4. Hereafter, we will only focus on the coupling gap in order to obtain the balanced OPS.

Likewise, $C1$, $C2$ and $C3$ are the coupling gap between WG1-WG2/WG3, WG2/WG3-WG4/WG5, WG4/WG5-WG6/WG7, respectively. In COMSOL simulations, the subdomains in the WG cross section were divided into triangular mesh elements. The grid size is set to $\lambda/15$ for the WGs geometries and $\lambda/8$ for the air domain so that precise simulation results can be obtained within the available computational resources. Scattering boundary conditions were applied at the outer edges of the FEM simulation window to estimate an open geometry.

Optimization of coupling gap, C1

The coupling strength can be very sensitive to the distance between the WGs and it is significant to ultimately pick a design that can function adequately given the type of deficiencies that are expected from the manufacturing process. In the first stage, $C1$ is varied from 30–200 nm between WG1 and the adjacent WGs (2 and 3). The coupling efficiency of $C1$ is plotted in fig. 2. It can be seen that when $C1 < 45$ nm, the energy doesn't transfer to the adjacent WGs instead it confines itself in $C1$. This happens because the gap is too small that it behaves as another slot WG and tries to confine the energy in $C1$. Therefore, an optimal separation between two WGs is important in order to obtain the maximum coupling.

A maximum coupling efficiency is obtained at values between 90–125 nm where the power is equally divided between WG2 and WG3 with less coupling losses. Hence, we used an optimal value of $C1 = 115$ nm in our splitter design for maximum coupling. Based on these results, $C2$ is also fixed at 115 nm to obtain maximum coupling between WGs (2, 3) and WGs (4, 5).

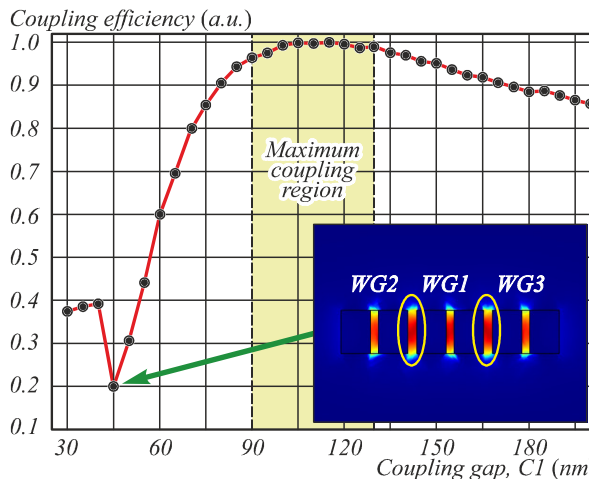


Fig. 2. Coupling efficiency $C1$ versus coupling gap

Optimization of coupling gap, C3

In order to obtain the balanced 1×4 OPS, $C3$ has to be slightly adjusted. If we consider on the right half of the power splitter, we can see that the energy from WG5 is transferred to WG7 and some part of the energy is coupled back to WG3. The distribution of the power between WG3 and WG7 is not equal as seen in fig. 3 where normalized intensity distribution at port 1, port 2, port 3 and port 4 are shown for various values of $C3$.

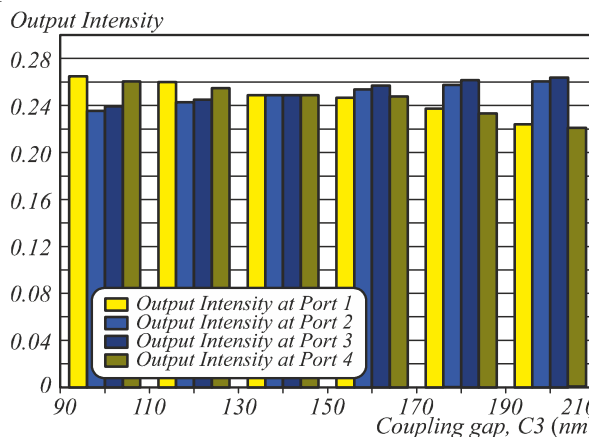


Fig. 3. Coupling gap, $C3$ versus intensity distribution at port 1, port 2, port 3 and port 4

For values of $C3 < 140$ nm, the output intensity at Port 2 and Port 3 is less than Port 1 and Port 4. For instance: at $C3 = 120$ nm, the intensity obtained at port 1, port 2, port 3 and port 4 is 26%, 24.2%, 24.4% and 25.4% respectively. Whereas on the other hand, when $C3$ increases, the intensity in Port 2 and Port 3 also increases gradually.

The optimized value of $C3$ is obtained at 140 nm where the OPS is balanced with 25% of power transmission at each port as shown in fig. 3. The line graph and the nor-

malized intensity contour plots at Port 1, Port 2, Port 3 and Port 4 are shown in fig. 4a and fig. 4b, respectively.

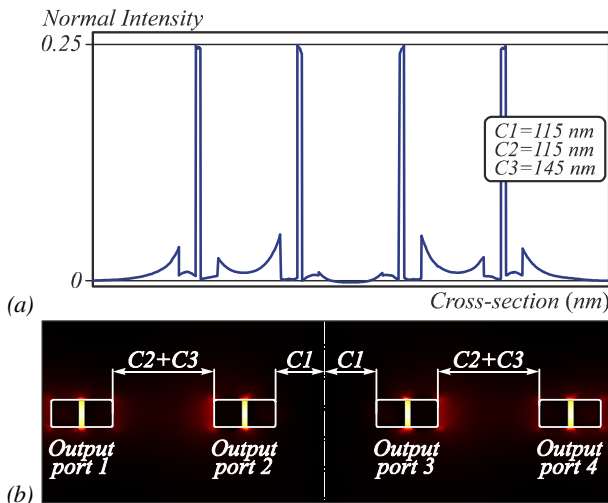


Fig. 4. Normalized intensity at Port 1, Port 2, Port 3 and Port 4, (a) Line graph, (b) Contour plot

Performance of an optimized balanced 1×4 OPS over whole C-band (1530 – 1565 nm)

In IR optical communications, C-band refers to the wavelength range 1530-1565 nm, which corresponds to the amplification range of erbium-doped fiber amplifier (EDFA). The 1×4 OPS was designed and optimized at 1550 nm. The performance of the power splitter was analysed and found that there is a maximum ~39% of power decay across the whole C-band as can be seen in fig 5. The parametric sweep function is used to calculate the electric field in the slot with a wavelength increment of 2. The designed power splitter remains balanced over the wavelength range of 1530–1565 nm.

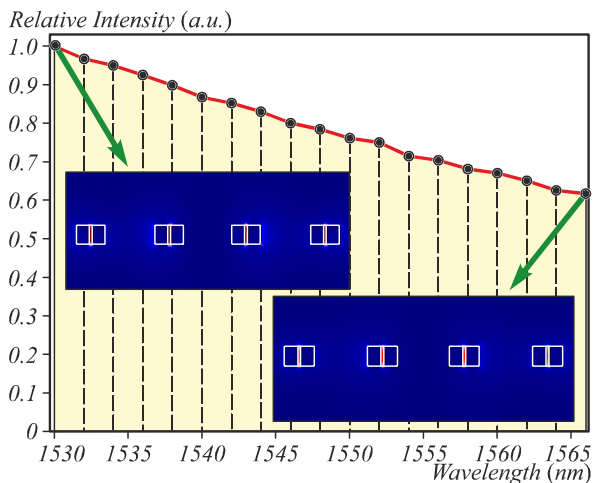


Fig. 5. Relative intensity at the output of the 1×4 OPS versus wavelength

Conclusion

In this paper, we presented a compact 1×4 balanced OPS based on silicon on insulator by using seven slot waveguides at TE polarization. This power splitter is capable of operating in whole infrared optical communication C-band (1530–1565 nm). Our designed splitter is a

promising candidate for high-density photonic integrated circuit due to its ultra-short length and excellent performance. Based on this design, splitters with various power distributions at the output can be established by varying the coupling gap. In future, we would like to extend this study to demonstrate similar OPS with a large number of output ports.

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Author's information

Muhammad Ali Butt, (b. 1985) received a Bachelor's degree in Electrical (Telecommunication) Engineering from Comsats Institute of Information and Technology, Pakistan in the year 2008. Then he left Pakistan for acquiring Master's degree from Germany. He attained Master's degree in Electrical Communication Engineering from University of Kassel (2010). He accomplished his PhD degree with Cum Laude in Material Sciences from Universitat Rovira i Virgili, Spain in the year 2015. In 2013, he made a research stay at Optoelectronic Research Centre (ORC), University of Southampton, England. Currently, he works as a Senior Scientist at Samara State Aerospace University, Russia. Research interests are optical waveguides, diffractive optics, and optical filters. E-mail: m.a.butt@ssau.ru.

Andra Naresh Kumar Reddy, received a Master degree in Physics (Electronics & Instrumentation) from University College of Science, Osmania University, India in the year 2005. He attained his Master of Philosophy in Physics from Bharatidasan University (2010). He completed his doctorate degree from Department of Physics, Osmania University, India in the year 2014. He worked as an Associate Professor of Physics in different autonomous engineering institutions in Hyderabad. He works as a Senior Scientist at Samara State Aerospace University, Russia. His research interests are diffractive optics, PSF engineering, thin film filters and metamaterials. E-mail: naareddy@gmail.com.

The information about author **Svetlana Nikolaevna Khonina** you can find on page 211 of this issue.

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